## SEMICONDUCTOR

## Features

- $2,048 \times 2,048$ channel non-blocking switching at 8.192 Mb/s
- Per-channel variable or constant throughput delay
- Automatic identification of ST-BUS/GCI interfaces
- Accept ST-BUS streams of $2.048 \mathrm{Mb} / \mathrm{s}, 4.096 \mathrm{Mb} / \mathrm{s}$ or $8.192 \mathrm{Mb} / \mathrm{s}$
- Automatic frame offset delay measurement
- Per-stream frame delay offset programming
- Per-channel high impedance output control
- Per-channel message mode
- Control interface compatible to Motorola non-mulitplexed CPUs
- Connection memory block programming
- IEEE-1149.1 (JTAG) Test Port


## Applications

- Medium and large switching platforms
- CTI application
- Voice/data multiplexer
- Digital cross connects
- ST-BUS/GCI interface functions
- Support IEEE 802.9a standard

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| :---: | :---: |
| Ordering Information |  |
| MT90820AP $\quad$ 84 Pin PLCC |  |
| MT90820AL | 100 Pin MQFP |
| -40 to $+85^{\circ} \mathrm{C}$ |  |

## Description

The MT90820 Large Digital Switch has a non-blocking switch capacity of $2,048 \times 2,048$ channels at a serial bit rate of $8.192 \mathrm{Mb} / \mathrm{s}, 1,024 \mathrm{x}$ 1,024 channels at $4.096 \mathrm{Mb} / \mathrm{s}$ and $512 \times 512$ channels at $2.048 \mathrm{Mb} / \mathrm{s}$. The device has many features that are programmable on per stream or per channel basis, including message mode, input offset delay and high impedance output control.

Per stream input delay control is particularly useful for managing large multi-chip switches that transport both voice channel and concatenated data channels.

In addition, input stream can be individually calibrated for input frame offset using a dedicated pin.


Figure 1 - Functional Block Diagram


Figure 2 - Pin Connections

Pin Description

| Pin \# |  | Name | Description |
| :---: | :---: | :---: | :---: |
| $\begin{gathered} \hline 84 \\ \text { PLCC } \end{gathered}$ | $\begin{gathered} 100 \\ \text { MQFP } \end{gathered}$ |  |  |
| $\begin{aligned} & 1,11, \\ & 30,54 \\ & 64,75 \end{aligned}$ | $\begin{aligned} & 31,41, \\ & 56,66, \\ & 76,99 \end{aligned}$ | $\mathrm{V}_{\text {SS }}$ | Ground. |
| $\begin{gathered} 2,32, \\ 63 \end{gathered}$ | $\begin{gathered} 5,40, \\ 67 \end{gathered}$ | $V_{D D}$ | +5 Volt Power Supply. |
| 3-10 | 68-75 | STo8-15 | ST-BUS Output 8 to 15 (Three-state Outputs): Serial data Output stream. These streams may have data rates of $2.048,4.096$ or $8.192 \mathrm{Mb} / \mathrm{s}$, depending upon the value programmed at bits DRO-1 in the IMS register. |
| $\begin{gathered} 12- \\ 27 \end{gathered}$ | 81-96 | STi0-15 | ST-BUS Input 0 to 15 (Inputs): Serial data input stream. These streams may have data rates of $2.048,4.096$ or $8.192 \mathrm{Mb} / \mathrm{s}$, depending upon the value programmed at bits DRO-1 in the IMS register. |
| 28 | 97 | F0i | Frame Pulse (Input): When the WFPS pin is low, this input accepts and automatically identifies frame synchronization signals formatted according to ST-BUS and GCI specifications. When the WFPS pin is high, this pin accepts a negative frame pulse which conforms to WFPS formats. |
| 29 | 98 | FE/HCLK | Frame Evaluation / HCLK Clock (Input): When the WFPS pin is low, this pin is the frame measurement input. When the WFPS pin is high, the HCLK (4.096MHz clock) is required for frame alignment in the wide frame pulse (WFP) mode. |
| 31 | 100 | CLK | Clock (Input): Serial clock for shifting data in/out on the serial streams (STi/o 015). Depending upon the value programmed at bits DRO-1 in the IMS register, this input accepts a $4.096,8.192$ or 16.384 MHz clock. |
| 33 | 6 | TMS | Test Mode Select (Input): JTAG signal that controls the state transitions of the TAP controller. This pin is pulled high by an internal pull-up when not driven. |
| 34 | 7 | TDI | Test Serial Data In (Input): JTAG serial test instructions and data are shifted in on this pin. This pin is pulled high by an internal pull-up when not driven. |
| 35 | 8 | TDO | Test Serial Data Out (Output): JTAG serial data is output on this pin on the falling edge of TCK. This pin is held in high impedance state when JTAG scan is not enable. |
| 36 | 9 | TCK | Test Clock (Input): Provides the clock to the JTAG test logic. This pin is pulled high by an internal pull-up when not driven. |
| 37 | 10 | $\overline{\text { TRST }}$ | Test Reset (Input): Asynchronously initializes the JTAG TAP controller by putting it in the Test-Logic-Reset state. This pin is pulled by an internal pull-up when not driven. This pin should be pulsed low on power-up, or held low, to ensure that the MT90820 is in the normal functional mode. |
| 38 | 11 | IC | Internal Connection (Input): Connect to $\mathrm{V}_{\mathrm{SS}}$ for normal operation. This pin must be low for the MT90820 to function normally and to comply with IEEE 1149 (JTAG) boundary scan requirements. This pin is pulled low internally when not driven. |
| 39 | 12 | RESET | Device Reset (Schmitt Trigger Input): This input (active LOW) puts the MT90820 in its reset state that clears the device internal counters, registers and brings STo0-15 and microport data outputs to a high impedance state. The time constant for a power up reset circuit must be a minimum of five times the rise time of the power supply. In normal operation, the RESET pin must be held low for a minimum of 100 nsec to reset the device. |

## Pin Description

| Pin \# |  | Name | Description |
| :---: | :---: | :---: | :---: |
| $\begin{array}{\|c\|} \hline 84 \\ \text { PLCC } \end{array}$ | $\begin{gathered} 100 \\ \text { MQFP } \end{gathered}$ |  |  |
| 40 | 13 | WFPS | Wide Frame Pulse Select (Input): When 1, enables the wide frame pulse (WFP) Frame Alignment interface. When 0, the device operates in ST-BUS/GCI mode. |
| $\begin{gathered} 41- \\ 48 \end{gathered}$ | 14-21 | A0-A7 | Address 0-7 (Input): When non-multiplexed CPU bus operation is selected, these lines provide the A0-A7 address lines to the internal memories. |
| 49 | 22 | DS/ $\overline{\mathrm{RD}}$ | Data Strobe / Read (Input): For multiplexed bus operation, this input is DS. This active high DS input works in conjunction with $\overline{\mathrm{CS}}$ to enable the read and write operations. <br> For Motorola non-multiplexed CPU bus operation, this input is DS. This active low input works in conjunction with $\overline{\mathrm{CS}}$ to enable the read and write operations. For multiplexed bus operation, this input is $\overline{\mathrm{RD}}$. This active low input sets the data bus lines (AD0-AD7, D8-D15) as outputs. |
| 50 | 23 | $\mathrm{R} / \overline{\mathrm{W}} / \overline{\mathrm{WR}}$ | Read/Write / Write (Input): In the cases of Motorola non-multiplexed and multiplexed bus operations, this input is R/W. This input controls the direction of the data bus lines (AD0-AD7, D8-D15) during a microprocessor access. For multiplexed bus operation, this input is WR. This active low input is used with $\overline{\mathrm{RD}}$ to control the data bus (ADO-7) lines as inputs. |
| 51 | 24 | $\overline{\mathrm{CS}}$ | Chip Select (Input): Active low input used by a microprocessor to activate the microprocessor port of MT90820. |
| 52 | 25 | AS/ALE | Address Strobe or Latch Enable (Input): This input is used if multiplexed bus operation is selected via the IM input pin. For Motorola non-multiplexed bus operation, connect this pin to ground. This pin is pulled low by an internal pull-down when not driven. |
| 53 | 26 | IM | CPU Interface Mode (input): When IM is high, the microprocessor port is in the multiplexed mode. When IM is low, the microprocessor port is in non-multiplexed mode. This pin is pulled low by an internal pull-down when not driven. |
| $\begin{gathered} 55- \\ 62 \end{gathered}$ | 32-39 | AD0-7 | Address/Data Bus 0 to 7 (Bidirectional): These pins are the eight least significant data bits of the microprocessor port. In multiplexed mode, these pins are also the input address bits of the microprocessor port. |
| $\begin{gathered} 65- \\ 72 \end{gathered}$ | 42-49 | D8-15 | Data Bus 8-15 (Bidirectional): These pins are the eight most significant data bits of the microprocessor port. |
| 73 | 50 | $\overline{\text { DTA }}$ | Data Transfer Acknowledgement (Active Low Output): Indicates that a data bus transfer is complete. When the bus cycle ends, this pin drives HIGH and then tri-states, allowing for faster bus cycles with a weaker pull-up resistor. A pull-up resistor is required to hold a HIGH level when the pin is tri-stated. |
| 74 | 55 | CSTo | Control Output (Output). This is a $4.096,8.192$ or $16.384 \mathrm{Mb} /$ s output containing 512,1024 or 2048 bits per frame respectively. The level of each bit is determined by the CSTo bit in the connection memory. See External Drive Control Section. |
| 76 | 57 | ODE | Output Drive Enable (Input): This is the output enable control for the SToO to STo15 serial outputs. When ODE input is low and the OSB bit of the IMS register is low, STo0-15 are in a high impedance state. If this input is high, the STo0-15 output drivers are enabled. However, each channel may still be put into a high impedance state by using the per channel control bit in the connection memory. |
| $\begin{gathered} 77- \\ 84 \end{gathered}$ | 58-65 | STo0-7 | Data Stream Output 0 to 7 (Three-state Outputs): Serial data Output stream. These streams have selectable data rates of $2.048,4.096$ or $8.192 \mathrm{Mb} / \mathrm{s}$. |

Pin Description

| Pin \# |  | Name | Description |
| :---: | :---: | :---: | :---: |
| $\begin{gathered} 84 \\ \text { PLCC } \end{gathered}$ | $\begin{aligned} & 100 \\ & \text { MQFP } \end{aligned}$ |  |  |
| - | $\begin{gathered} 1-4, \\ 27-30, \\ 51-54 \\ 77-80 \end{gathered}$ | NC | No connection. |

## Device Overview

The MT90820 Large Digital Switch is capable of switching up to $2,048 \times 2,048$ channels. The MT90820 is designed to switch 64 kbit/s PCM or N x $64 \mathrm{kbit} / \mathrm{s}$ data. The device maintains frame integrity in data applications and minimum throughput delay for voice applications on a per channel basis.

The serial input streams of the MT90820 can have a bit rate of $2.048,4.096$ or $8.192 \mathrm{Mbit} / \mathrm{s}$ and are arranged in $125 \mu \mathrm{~s}$ wide frames, which contain 32,64 or 128 channels, respectively. The data rates on input and output streams are identical.

By using Mitel's message mode capability, the microprocessor can access input and output time-slots on a per channel basis. This feature is useful for transferring control and status information for external circuits or other ST-BUS devices. The MT90820 automatically identifies the polarity of the frame synchronization input signal and configures its serial streams to be compatible to either ST-BUS or GCI formats.

Two different microprocessor bus interfaces can be selected through the Input Mode pin (IM): Non-multiplexed or Multiplexed. These interfaces provide compatibility with multiplexed and Motorola non-multiplexed buses.

The frame offset calibration function allows users to measure the frame offset delay using a frame evaluation pin (FE). The input offset delay can be programmed for individual streams using internal frame input offset registers, see Table 11.

The internal loopback allows the ST-BUS output data to be looped around to the ST-BUS inputs for diagnostic purposes.

## Functional Description

A functional Block Diagram of the MT90820 is shown in Figure 1.

## Data and Connection Memory

For all data rates, the received serial data is converted to parallel format by internal serial-to-parallel converters and stored sequentially in the data memory. Depending upon the selected operation programmed in the interface mode select (IMS) register, the useable data memory may be as large as 2,048 bytes. The sequential addressing of the data memory is performed by an internal counter, which is reset by the input 8 kHz frame pulse (F0i) to mark the frame boundaries of the incoming serial data streams.

Data to be output on the serial streams may come from either the data memory or connection memory. Locations in the connection memory are associated with particular ST-BUS output channels. When a channel is due to be transmitted on an ST-BUS output, the data for this channel can be switched either from an ST-BUS input in connection mode, or from the lower half of the connection memory in message mode. Data destined for a particular channel on a serial output stream is read from the data memory or connection memory during the previous channel time-slot. This allows enough time for memory access and parallel-to-serial conversion.

## Connection and Message Modes

In the connection mode, the addresses of the input source data for all output channels are stored in the connection memory. The connection memory is mapped in such a way that each location corresponds to an output channel on the output streams. For details on the use of the source address data (CAB and SAB bits), see Table 13 and Table 14. Once the source address bits are programmed by the microprocessor, the contents of
the data memory at the selected address are transferred to the parallel-to-serial converters and then onto an ST-BUS output stream.

By having several output channels connected to the same input source channel, data can be broadcasted from one input channel to several output channels.

In message mode, the microprocessor writes data to the connection memory locations corresponding to the output stream and channel number. The lower half ( 8 least significant bits) of the connection memory content is transferred directly to the parallel-to-serial converter. This data will be output on the ST-BUS streams in every frame until the data is changed by the microprocessor.

The five most significant bits of the connection memory controls the following for an output channel: message or connection mode, constant or variable delay, enables/tristate the ST-BUS output drivers and enables/disable the loopback function. In addition, one of these bits allows the user to control the CSTo output.

If an output channel is set to a high-impedance state through the connection memory, the ST-BUS output will be in a high impedance state for the duration of that channel. In addition to the per-channel control, all channels on the ST-BUS outputs can be placed in a high impedance state by either pulling the ODE input pin low or programming the output stand by (OSB) bit in the interface mode selection register to low. This action overrides the individual per-channel programming by the connection memory bits.

The connection memory data can be accessed via the microprocessor interface through the D0 to D15 pins. The addressing of the device internal registers, data and connection memories is performed through the address input pins and the Memory Select (MS) bit of the control register. For details on device addressing, see Software Control and Control Register bits description (Table 4, Tables 6 and 7).

## Serial Data Interface Timing

The master clock frequency must always be twice the data rate. The master clock (CLK) must be either at $4.096,8.192$ or 16.384 MHz for serial data rate of $2.048,4.096$ or $8.192 \mathrm{Mb} / \mathrm{s}$ respectively. The input and output stream data rates will always be identical.

The MT90820 provides two different interface timing modes controlled by the WFPS pin. If the WFPS pin is low, the MT90820 is in ST-BUS/GCI mode. If the

WFPS pin is high, the MT90820 is in the wide frame pulse (WFP) frame alignment mode.

In ST-BUS/GCI mode, the input 8 kHz frame pulse can be in either ST-BUS or GCl format. The MT90820 automatically detects the presence of an input frame pulse and identifies it as either ST-BUS or GCI. In ST-BUS format, every second falling edge of the master clock marks a bit boundary and the data is clocked in on the rising edge of CLK, three quarters of the way into the bit cell, see. In GCl format, every second rising edge of the master clock marks the bit boundary and data is clocked in on the falling edge of CLK at three quarters of the way into the bit cell, see Figure12.

## Wide Frame Pulse (WFP) Frame Alignment Timing

When the device is in WFP frame alignment mode, the CLK input must be at 16.384 MHz , the FE/HCLK input is 4.096 MHz and the 8 kHz frame pulse is in ST-BUS format. The timing relationship between CLK, HCLK and the frame pulse is defined in Figure 13.

When WFPS pin is high, the frame alignment evaluation feature is disabled, but the frame input offset registers may still be programmed to compensate for the varying frame delays on the serial input streams.

## Switching Configurations

The MT90820 maximum non-blocking switching configurations is determined by the data rates selected for the serial inputs and outputs. The switching configuration is selected by two DR bits in the IMS register. See Table 8 and Table 9.

## 

When the $2.048 \mathrm{Mb} / \mathrm{s}$ data rate is selected, the device is configured with 16-input/16-output data streams each having 32 64Kbit/s channels each. This mode requires a CLK of 4.094 MHz and allows a maximum non-blocking capacity of 512 x 512 channels.

## $4.096 \mathrm{Mb} / \mathrm{s}$ Serial Links $(D R 0=1, D R 1=0)$

When the $4.096 \mathrm{Mb} / \mathrm{s}$ data rate is selected, the device is configured with 16 -input/16-output data streams each having 64 64Kbit/s channels each. This mode requires a CLK of 8.192 MHz and allows a maximum non-blocking capacity of $1,024 \times 1,024$ channels.
$8.192 \mathrm{Mb} / \mathrm{s}$ Serial Links (DR0=0, DR1=1)
When the $8.192 \mathrm{Mb} / \mathrm{s}$ data rate is selected, the device is configured with 16 -input/16-output data streams each having 128 64Kbit/s channels each. This mode requires a CLK of 16.384 MHz and allows a maximum non-blocking capacity of $2,048 \times 2,048$ channels. Table 1 summarizes the switching configurations and the relationship between different serial data rates and the master clock frequencies.

| Serial <br> Interface <br> Data Rate | Master Clock <br> Required <br> $(M H z)$ | Matrix <br> Channel <br> Capacity |
| :---: | :---: | :---: |
| $2 \mathrm{Mb} / \mathrm{s}$ | 4.096 | $512 \times 512$ |
| $4 \mathrm{Mb} / \mathrm{s}$ | 8.192 | $1,024 \times 1,024$ |
| $8 \mathrm{Mb} / \mathrm{s}$ | 16.384 | $2,048 \times 2,048$ |

Table 1-Switching Configuration

## Input Frame Offset Selection

Input frame offset selection allows the channel alignment of individual input streams to be offset with respect to the output stream channel alignment (i.e. FOi). This feature is useful in compensating for variable path delays caused by serial backplanes of variable lengths, which may be implemented in large centralized and distributed switching systems.

Each input stream can have its own delay offset value by programming the frame input offset (FOR) registers. Possible adjustment can range up to +4 master clock (CLK) periods forward with resolution of $1 / 2$ clock period. The output frame offset cannot be offset or adjusted. See Figure 4, Table 11 and Table 12 for delay offset programming.

## Serial Input Frame Alignment Evaluation

The MT90820 provides the frame evaluation (FE) input to determine different data input delays with respect to the frame pulse F0i.

A measurement cycle is started by setting the start frame evaluation (SFE) bit low for at least one frame. Then the evaluation starts when the SFE bit in the IMS register is changed from low to high. Two frames later, the complete frame evaluation (CFE) bit of the frame alignment register (FAR) changes from low to high to signal that a valid offset measurement is ready to be read from bits 0 to 11 of the FAR register. The SFE bit must be set to zero before a new measurement cycle started.

In ST-BUS mode, the falling edge of the frame measurement signal (FE) is evaluated against the falling edge of the ST-BUS frame pulse. In GCI mode, the rising edge of FE is evaluated against the rising edge of the GCI frame pulse. See Table 10 \& Figure 3 for the description of the frame alignment register.

This feature is not available when the WFP Frame Alignment mode is enabled (i.e., when the WFPS pin is connected to VDD).

## Memory Block Programming

The MT90820 provides users with the capability of initializing the entire connection memory block in two frames. Bits 11 to 15 of every connection memory location will be programmed with the pattern stored in bits 5 to 9 of the IMS register.

The block programming mode is enabled by setting the memory block program (MBP) bit of the control register high. When the block programming enable (BPE) bit of the IMS register is set to high, the block programming data will be loaded into the bits 11 to 15 of every connection memory location. The other connection memory bits (bit 0 to bit 10) are loaded with zeros. When the memory block programming is complete, the device resets the BPE bit to zero.

## Loopback Control

The loopback control (LPBK) bit of each connection memory location allows the ST-BUS output data to be looped backed internally to the ST-BUS input for diagnostic purposes.

If the LPBK bit is high, the associated ST-BUS output channel data is internally looped back to the ST-BUS input channel (i.e., data from STo $n$ channel $m$ routes to the STi $n$ channel $m$ internally); if the LPBK bit is low, the loopback feature is disabled. For proper per-channel loopback operation, the contents of frame delay offset registers must be set to zero.

## Delay Through the MT90820

The switching of information from the input serial streams to the output serial streams results in a throughput delay. The device can be programmed to perform time-slot interchange functions with different throughput delay capabilities on the per-channel basis. For voice application, select variable throughput delay to ensure minimum delay between input and output data. In wideband data applications,
select constant throughput delay to maintain the frame integrity of the information through the switch.

The delay through the device varies according to the type of throughput delay selected in the $\bar{\nabla} / C$ bit of the connection memory.

## Variable Delay Mode ( $\overline{\mathrm{V}} / \mathrm{C}$ bit $=0$ )

The delay in this mode is dependent only on the combination of source and destination channels and is independent of input and output streams. The minimum delay achievable in the MT90820 is three time-slots. When the input channel data is switched to the same output channel (channel $n$, frame p ), it will be output in the following frame (channel $n$, frame $p+1$ ). The same frame delay occurs if the input channel n is switched to output channel $\mathrm{n}+1$ or $\mathrm{n}+2$. When input channel n is switched to output channel $n+3, n+4, \ldots$, the new output data will appear in the same frame. Table 2 shows the possible delays for the MT90820 in the variable delay mode.

## Constant Delay Mode ( $\overline{\mathrm{V}} / \mathrm{C}$ bit $=1$ )

In this mode, frame integrity is maintained in all switching configurations by making use of a multiple data memory buffer. Input channel data is written into the data memory buffers during frame $n$ will be read out during frame $n+2$.

In the MT90820, the minimum throughput delay achievable in the constant delay mode will be one frame. For example, in $2 \mathrm{Mb} / \mathrm{s}$ mode, when input time-slot 31 is switched to output time-slot 0 . The maximum delay of 94 time-slots of delay occurs when time-slot 0 in a frame is switched to time-slot 31 in the frame. See Table 3.

## Microprocessor Interface

The MT90820 provides a parallel microprocessor interface for non-multiplexed or multiplexed bus structures. This interface is compatible with Motorola non-multiplexed and multi-plexed buses.

If the IM pin is low, the MT90820 microprocessor interface assumes Motorola non-multiplexed bus mode. If the IM pin is high, the device micro-processor interface accepts two different timing modes (mode1 and mode2) which allows direct connection to multiplexed microprocessors.

The microprocessor interface automatically identifies the type of micro-processor bus connected to the

MT90820. This circuit uses the level of the DS/ $\overline{R D}$ input pin at the rising edge of AS/ALE to identify the appropriate bus timing connected to the MT90820. If $D S / \overline{R D}$ is low at the rising edge of AS/ALE, then the mode 1 multiplexed timing is selected. If $\mathrm{DS} / \overline{\mathrm{RD}}$ is high at the rising edge of AS/ALE, then the mode 2 multiplexed bus timing is selected.

For multiplexed operation, the required signals are the 8 -bit data and address (AD0-AD7), 8-bit Data (D8-D15), Address strobe/Address latch enable (AS/ ALE), Data strobe/Read (DS/RD), Read/Write /Write ( $\mathrm{R} / \overline{\mathrm{W}} / \overline{\mathrm{WR}}$ ), Chip select ( $\overline{\mathrm{CS}}$ ) and Data transfer acknowledge ( $\overline{\mathrm{DTA}}$ ). See Figure 13 and Figure 14 for multiplexed parallel microport timing.

For the Motorola non-multiplexed bus, the required signals are the 16-bit data bus (AD0-AD7, D8-D15), 8 -bit address bus (A0-A7) and 4 control lines ( $\overline{C S}$, DS, R/W and DTA). See Figure 15 for Motorola non-multiplexed microport timing.

The MT90820 microport provides access to the internal registers, connection and data memories. All locations provide read/write access except for the data memory and the frame alignment register which are read only.

## Memory Mapping

The address bus on the microprocessor interface selects the internal registers and memories of the MT90820. If the A7 address input is low, then the control (CR), interface mode selection (IMS), frame alignment (FAR) and frame input offset (FOR) registers are addressed by A6 to A0 according to Table 4.
If the A7 is high, then the remaining address input lines are used to select memory subsections of up to 128 locations corresponding to the maximum number of channels per input or output stream. The address input lines and the stream address bits (STA) of the control register allow access to the entire data and connection memories.

The control and IMS registers together control all the major functions of the device. The IMS register should be programmed immediately after system power-up to establish the desired switching configuration as explained in the Serial Data Interface Timing and Switching Configurations sections.

The control register is used to control switching operations in the MT90820. It selects the internal memory locations that specify the input and output channels selected for switching.

The data in the control register consists of the memory block programming bit (MBP), the memory select bit (MS) and the stream address bits (STA). The memory block programming bit allows users to program the entire connection memory block, (see Memory Block Programming section). The memory select bit controls the selection of the connection memory or the data Memory. The stream address bits define an internal memory subsections corresponding to input or output ST-BUS streams.

The data in the IMS register consists of block programming bits (BPD0-BPD4), block program-ming enable bit (BPE), output stand by bit (OSB), start frame evaluation bit (SFE) and data rate selection bits (DR0, DR1). The block programming and the block programming enable bits allows users to program the entire connection memory, (see Memory Block Programming section). If the ODE pin is low, the OSB bit enables (if high) or disables (if low) all ST-BUS output drivers. If the ODE pin is high, the contents of the OSB bit is ignored and all ST-BUS output drivers are enabled.

## Connection Memory Control

The contents of the CSTo bit of each connection memory location are output on the CSTo pin once every frame. The CSTo pin is a 4.096, 8.192 or $16.384 \mathrm{Mb} / \mathrm{s}$ output, which carries $512,1,024$ or 2,048 bits, respectively. If the CSTo bit is set high, the corresponding bit on the CSTo output is transmitted high. If the CSTo bit is low, the corresponding bit on the CSTo output is transmitted
low. The contents of the CSTo bits of the connection memory are transmitted sequentially on to the CSTo pin and are synchronous with the data rates on the other ST-BUS streams.

The CSTO bit is output one channel before the corresponding channel on the ST-BUS. For example, in $2 \mathrm{Mb} / \mathrm{s}$ mode, the contents of the CSTo bit in position 0 (SToO, CHO) of the connection memory is output on the first clock cycle of channel 31 through CSTo pin. The contents of the CSTo bit in position 32 (STo1, CHO) of the connection memory is output on the second clock cycle of channel 31 via CSTo pin.

If the ODE pin or the OSB bit is high, the OE bit of each connection memory location enables (if high) or disables (if low) the output drivers for an individual ST-BUS output stream and channel. See Table 5 for detail.

The message channel (MC) bit of the connection memory enables (if high) an associated ST-BUS output channel in message mode. If the MC bit is low, the contents of the stream address bit (SAB) and the channel address bit (CAB) of the connection memory defines the source information (stream and channel) of the time-slot that will be switched to the output. When message mode is enabled, only the lower half (8 least significant bits) of the connection memory is transferred to the ST-BUS outputs.

Bit $\overline{\mathrm{V}} / \mathrm{C}$ (Variable/Constant Delay) of each connection memory location allows the per-channel selection

| Input Rate | Delay for Variable Throughput Delay Mode ( $m$ - output channel number) ( n - input channel number)) |  |  |
| :---: | :---: | :---: | :---: |
|  | $\mathrm{m}<\mathrm{n}$ | $\mathrm{m}=\mathrm{n}, \mathrm{n}+1, \mathrm{n}+2$ | $\mathrm{m}>\mathrm{n}+2$ |
| $2.048 \mathrm{Mb} / \mathrm{s}$ | $32-(\mathrm{n}-\mathrm{m})$ time-slots | $\mathrm{m}-\mathrm{n}+32$ time-slots | m -n time-slots |
| $4.096 \mathrm{Mb} / \mathrm{s}$ | 64 - (n-m) time-slots | $\mathrm{m}-\mathrm{n}+64$ time-slots | m-n time-slots |
| 8.192 Mb/s | 128-(n-m) time-slots | $m-n+128$ time-slots | m-n time-slots |

Table 2 - Variable Throughput Delay Value

| Input Rate | Delay for Constant Throughput Delay Mode <br> $(\mathbf{m}$ - output channel number) <br> $(\mathbf{n}-$ input channel number)) |
| :---: | :---: |
| $2.048 \mathrm{Mb} / \mathrm{s}$ | $32+(32-\mathrm{n})+(\mathrm{m}-1)$ time-slots |
| $4.096 \mathrm{Mb} / \mathrm{s}$ | $64+(64-\mathrm{n})+(\mathrm{m}-1)$ time-slots |
| $8.192 \mathrm{Mb} / \mathrm{s}$ | $128+(128-\mathrm{n})+(\mathrm{m}-1)$ time-slots |

Table 3-Constant Throughput Delay Value
between variable and constant throughput delay modes.

If the LPBK bit is high, the associated ST-BUS output channel data is internally looped back to the ST-BUS input channel (i.e., STi $n$ channel $m$ data comes from the STo $n$ channel $m$ ). If the LPBK bit is low, the loopback feature is disabled. For proper per-channel loopback operation, the contents of the frame delay offset registers must be set to zero.

During the microprocessor initialization routine, the microprocessor should program the desired active paths through the switch, and put all other channels into a high impedance state. This procedure prevents two ST-BUS outputs from driving the same stream simultaneously. When this process is complete, the microprocessor controlling the matrices can bring the ODE pin or OSB bit high to relinquish the high impedance state control to the OE bit in the connection memory.

## Initialization of the MT90820

After power up, the contents of the connection memory can be in any state. The ODE pin should be held low after power up to keep all ST-BUS outputs in a high impedance state until the microprocessor has initialized the switching matrix.

| A7 <br> (Note 1) | A6 | A5 | A4 | A3 | A2 | A1 | A0 | Location |
| :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: |
| 0 | 0 | 0 | 0 | 0 | 0 | 0 | 0 | Control Register, CR |
| 0 | 0 | 0 | 0 | 0 | 0 | 0 | 1 | Interface Mode Selection Register, IMS |
| 0 | 0 | 0 | 0 | 0 | 0 | 1 | 0 | Frame Alignment Register, FAR |
| 0 | 0 | 0 | 0 | 0 | 0 | 1 | 1 | Frame Input Offset Register 0, FOR0 |
| 0 | 0 | 0 | 0 | 0 | 1 | 0 | 0 | Frame Input Offset Register 1, FOR1 |
| 0 | 0 | 0 | 0 | 0 | 1 | 0 | 1 | Frame Input Offset Register 2, FOR2 |
| 0 | 0 | 0 | 0 | 0 | 1 | 1 | 0 | Frame Input Offset Register 3, FOR3 |
| 1 | 0 | 0 | 0 | 0 | 0 | 0 | 0 | Ch 0 |
| 1 | 0 | 0 | 0 | 0 | 0 | 0 | 1 | Ch 1 |
| 1 | 0 | 0 |  |  |  |  |  |  |
| 1 | 0 | 0 | 1 | 1 | 1 | 1 | 0 | Ch 30 |
| 1 | 0 | 0 | 1 | 1 | 1 | 1 | 1 | Ch 31 (Note 2) |
| 1 | 0 | 1 | 0 | 0 | 0 | 0 | 0 | Ch 32 |
| 1 | 0 | 1 | 0 | 0 | 0 | 0 | 1 | Ch 33 |
| 1 | 0 | 1 |  |  |  |  |  |  |
| 1 | 0 | 1 | 1 | 1 | 1 | 1 | 0 | Ch 62 |
| 1 | 0 | 1 | 1 | 1 | 1 | 1 | 1 | Ch 63 (Note 3) |
| 1 | 1 | 0 | 0 | 0 | 0 | 0 | 0 | Ch 64 |
| 1 | 1 | 0 | 0 | 0 | 0 | 0 | 1 | Ch 65 |
| 1 | 1 |  |  |  |  |  |  |  |
| 1 | 1 | 1 | 1 | 1 | 1 | 1 | 0 | Ch 126 |
| 1 | 1 | 1 | 1 | 1 | 1 | 1 | 1 | Ch 127 (Note 4) |
| Notes: <br> 1. Bit A7 must be high for access to data and connection memory positions. Bit A7 must be low for access to registers. <br> 2. Channels 0 to 31 are used when serial interface is at $2 \mathrm{Mb} / \mathrm{s}$ mode. <br> 3. Channels 0 to 63 are used when serial interface is at $4 \mathrm{Mb} / \mathrm{s}$ mode. <br> 4. Channels 0 to 127 are used when serial interface is at $8 \mathrm{Mb} / \mathrm{s}$ mode. |  |  |  |  |  |  |  |  |

Table 4 - Internal Register and Address Memory Mapping

| OE bit in Connection <br> Memory | ODE pin | OSB bit in IMS <br> register | ST-BUS Output Driver <br> Status |
| :---: | :---: | :---: | :---: |
| 0 | Don't Care | Don't Care | Per Channel <br> High Impedance |
| 1 | 0 | 0 | High Impedance |
| 1 | 0 | 1 | Enable |
| 1 | 1 | 0 | Enable |
| 1 | 1 | 1 | Enable |

Table 5 - Output High Impedance Control


Table 6 - Control (CR) Register Bits

| Input/Output <br> Data Rate | Valid Address Lines |
| :---: | :---: |
| $2.048 \mathrm{Mb} / \mathrm{s}$ | $\mathrm{A} 4, \mathrm{~A} 3, \mathrm{~A} 2, \mathrm{~A} 1, \mathrm{~A} 0$ |
| $4.096 \mathrm{Mb} / \mathrm{s}$ | $\mathrm{A} 5, \mathrm{~A} 4, \mathrm{~A} 3, \mathrm{~A} 2, \mathrm{~A} 1, \mathrm{~A} 0$ |
| $8.192 \mathrm{Mb} / \mathrm{s}$ | $\mathrm{A} 6, \mathrm{~A} 5, \mathrm{~A} 4 \mathrm{~A} 3, \mathrm{~A} 2, \mathrm{~A} 1, \mathrm{~A} 0$ |

Table 7 - Valid Address lines for Different Bit Rates


Table 8 - Interface Mode Selection (IMS) Register Bits

| DR1 | DR0 | Data Rate Selected | Master Clock Required |
| :---: | :---: | :---: | :---: |
| 0 | 0 | $2.048 \mathrm{Mb} / \mathrm{s}$ | 4.096 MHz |
| 0 | 1 | $4.096 \mathrm{Mb} / \mathrm{s}$ | 8.192 MHz |
| 1 | 0 | $8.192 \mathrm{Mb} / \mathrm{s}$ | 16.384 MHz |
| 1 | 1 | Reserved | Reserved |

Table 9 - Serial Data Rate Selection (16 input x 16 output)


Table 10 - Frame Alignment (FAR) Register Bits


Figure 3 - Example for Frame Alignment Measurement

Read/Write Address: $\quad 03_{\mathrm{H}}$ for FOR0 register, $04_{\mathrm{H}}$ for FOR1 register, $05_{\mathrm{H}}$ for FOR2 register, $06_{H}$ for FOR3 register,
Reset value: $\quad 0000_{\mathrm{H}}$ for all FOR registers.


FOR0 register

| 15 |
| :---: |
| 14 |
| 14 |
| 13 |
| 12 |

FOR1 register

| 15 | 14 | 13 | 12 | 11 | 10 | 9 | 8 | 7 | 6 | 5 | 4 | 3 | 2 | 1 | 0 |
| :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: |
| OF112 | OF111 | OF110 | DLE11 | OF102 | OF101 | OF100 | DLE10 | OF92 | OF91 | OF90 | DLE9 | OF82 | OF81 | OF80 | DLE8 |

FOR2 register

| 15 | 14 | 13 | 12 | 11 | 10 | 9 | 8 | 7 | 6 | 5 | 4 | 3 | 2 | 1 | 0 |
| :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: |
| OF152 | OF151 | OF150 | DLE15 | OF142 | OF141 | OF140 | DLE14 | OF132 | OF131 | OF130 | DLE13 | OF122 | OF121 | OF120 | DLE12 |

FOR3 register

| Name (Note 1) | Description |
| :---: | :---: |
| OFn2, OFn1, OFn0 | Offset Bits $2,1 \& 0$. These three bits define how long the serial interface receiver takes to recognize and store bit 0 from the STi input pin: i.e., to start a new frame. The input frame offset can be selected to +4 clock periods from the point where the external frame pulse input signal is applied to the FOi input of the device. See Figure 4. |
| DLEn | Data Latch Edge. <br> ST-BUS mode: DLEn $=0$, if clock rising edge is at the $3 / 4$ point of the bit cell. DLEn $=1$, if when clock falling edge is at the $3 / 4$ of the bit cell. <br> GCI mode: $\quad$ DLEn $=0$, if clock falling edge is at the $3 / 4$ point of the bit cell. <br> DLEn $=1$, if when clock rising edge is at the $3 / 4$ of the bit cell. |

Note 1: n denotes an input stream number from 0 to 15.
Table 11 - Frame Input Offset (FOR) Register Bits

| Input Stream <br> Offset | Measurement Result from <br> Frame Delay Bits |  |  |  | Corresponding <br> Offset Bits |  |  |  |  |
| :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: |
|  | FD11 | FD2 | FD1 | FD0 | OFn2 | OFn1 | OFn0 | DLEn |  |
| No clock period shift (Default) | 1 | 0 | 0 | 0 | 0 | 0 | 0 | 0 |  |
| + 0.5 clock period shift | 0 | 0 | 0 | 0 | 0 | 0 | 0 | 1 |  |
| +1.0 clock period shift | 1 | 0 | 0 | 1 | 0 | 0 | 1 | 0 |  |
| +1.5 clock period shift | 0 | 0 | 0 | 1 | 0 | 0 | 1 | 1 |  |
| +2.0 clock period shift | 1 | 0 | 1 | 0 | 0 | 1 | 0 | 0 |  |
| +2.5 clock period shift | 0 | 0 | 1 | 0 | 0 | 1 | 0 | 1 |  |
| +3.0 clock period shift | 1 | 0 | 1 | 1 | 0 | 1 | 1 | 0 |  |
| +3.5 clock period shift | 0 | 0 | 1 | 1 | 0 | 1 | 1 | 1 |  |
| +4.0 clock period shift | 1 | 1 | 0 | 0 | 1 | 0 | 0 | 0 |  |
| +4.5 clock period shift | 0 | 1 | 0 | 0 | 1 | 0 | 0 | 1 |  |

Table 12-Offset Bits (OFn2, OFn1, OFn0, DLEn) \& Frame Delay Bits (FD11, FD2-0)


Figure 4 - Examples for Input Offset Delay Timing

| 15 | 14 | 13 | 12 | 11 | 10 | 9 | 8 | 7 | 6 | 5 | 4 | 3 | 2 | 1 | 0 |
| :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: |
| LPBK | V/C | MC | CSTo | OE | SAB3 | SAB2 | SAB1 | SABO | CAB6 | CAB5 | CAB4 | CAB3 | CAB2 | CAB1 | CABO |
| Bit |  | Nam |  |  |  |  |  |  |  | escrip | tion |  |  |  |  |
| 15 |  | LPB |  |  | Per Chan STo n ch offset reg mode. | nnel L annel ister b |  | ack. W prope n[2:0] | hen 1 , per o zero | the S hannel for th | in ch loopb strea | nnel ack op ms wh | m data eration ich are | come <br> s, se <br> in th | from the the delay loopback |
| 14 |  | V/C |  |  | Variable the varia basis. | /Cons <br> ble (low) | tant T <br> w) and | hrou the | hput nstan | Delay. delay | This b (high) | is us mode | d to s on a | elect per-ch | etween annel |
| 13 |  | MC |  |  | Message <br> output on (bit 7 - bit the conn channel | Chan <br> the c 0) will ection and str | orresp be ou memory eam. | When 1 onding tput to ry are | , the c outpu the S the da | ontent than -BUS ta mem | of th nel and outpu nory | conn strea pins. ddress | ection <br> m. On When 0 of the | memo y the , the c switch | ry are ower byte ontents of ed input |
| 12 |  | CS |  |  | Control The CS | ST-BU <br> o bit fo | S outp strea | $\begin{aligned} & \text { sut. } \mathrm{Tr} \\ & \text { am } 0 \text { is } \end{aligned}$ | is bit is outpu | outpu first. | on th | $\mathrm{e} \text { CST }$ | pin or | ne cha | nnel early. |
| 11 |  | OE |  |  | Output basis. When 1, high-imp | Enable <br> the ou edance | . This <br> tput drive state | bit ena <br> iver fu | bles th <br> ctions | norm |  | tput <br> en 0 | ivers <br> the ou | tput dr | r-channel ver is in a |
| $\begin{gathered} 10-8 \\ 7 \\ \text { (Note 1) } \end{gathered}$ |  | SAB |  |  | Source <br> stream f | Stream <br> or the | Add source | ess of th | ts. Th conne | e bina ction. | y valu | is th | numb | er of | he data |
| $\begin{gathered} 6-0 \\ \text { (Note 1) } \end{gathered}$ |  | CAB |  |  | Source channel | Chann for the |  | dress <br> of th | Bits. T conn | he bin ection | ary va | e is | he num | ber | the |
| Note 1: If bit $13(\mathrm{MC})$ of the corresponding connection memory location is 1 (device in message mode), then these entire 8 bits (SAB0, CAB6-CAB0) are output on the output channel and stream associated with this location. |  |  |  |  |  |  |  |  |  |  |  |  |  |  |  |

Table 13 - Connection Memory Bits

| Data Rate | CAB Bits Used to Determine the Source Channel of the Connection |
| :---: | :---: |
| $2.048 \mathrm{Mb} / \mathrm{s}$ | CAB4 to CAB0 (32 channel/input stream) |
| $4.096 \mathrm{Mb} / \mathrm{s}$ | CAB5 to CAB0 (64 channel/input stream) |
| $8.192 \mathrm{Mb} / \mathrm{s}$ | CAB6 to CAB0 (128 channel/input stream) |

Table 14-CAB Bits Programming for Different Data Rates

## JTAG Support

The MT90820 JTAG interface conforms to the Boundary-Scan standard IEEE1149.1. This standard specifies a design-for-testability technique called Boundary-Scan test (BST). The operation of the boundary-scan circuitry is controlled by an external test access port (TAP) Controller.

## Test Access Port (TAP)

The Test Access Port (TAP) provides access to the many test functions of the MT90820. It consists of three input pins and one output pin. The following pins are from the TAP.

- Test Clock Input (TCK)

TCK provides the clock for the test logic. The TCK does not interfere with any on-chip clock and thus remain independent. The TCK permits shifting of test data into or out of the Boundary-Scan register cells concurrently with the operation of the device and without interfering with the on-chip logic.

- Test Mode Select Input (TMS) The logic signals received at the TMS input are interpreted by the TAP Controller to control the test operations. The TMS signals are sampled at the rising edge of the TCK pulse. This pin is internally pulled to Vdd when it is not driven from an external source.
- Test Data Input (TDI)

Serial input data applied to this port is fed either into the instruction register or into a test data register, depending on the sequence previously applied to the TMS input. Both registers are described in a subsequent section. The received input data is sampled at the rising edge of TCK pulses. This pin is internally pulled to Vdd when it is not driven from an external source.

- Test Data Output (TDO)

Depending on the sequence previously applied to the TMS input, the contents of either the instruction register or data register are serially shifted out towards the TDO. The data out of the TDO is clocked on the falling edge of the TCK pulses. When no data is shifted through the boundary scan cells, the TDO driver is set to a high impedance state.

- Test Reset (TRST)

Reset the JTAG scan structure. This pin is internally pulled to VDD.

## Instruction Register

In accordance with the IEEE 1149.1 standard, the MT90820 uses public instructions. The MT90820 JTAG Interface contains a two-bit instruction register. Instructions are serially loaded into the instruction register from the TDI when the TAP Controller is in its shifted-IR state. Subsequently, the instructions are decoded to achieve two basic functions: to select the test data register that may operate while the instruction is current, and to define the serial test data register path, which is used to shift data between TDI and DO during data register scanning.

## Test Data Register

As specified in IEEE 1149.1, the MT90820 JTAG Interface contains two test data registers:

- The Boundary-Scan register The Boundary-Scan register consists of a series of Boundary-Scan cells arranged to form a scan path around the boundary of the MT90820 core logic.
- The Bypass Register The Bypass register is a single stage shift register that provides a one-bit path from TDI to its TDO.

The MT90820 boundary scan register contains 118 bits. Bit 0 in Table 15 Boundary Scan Register is the first bit clocked out. All tristate enable bits are active high.

| Device Pin | Boundary Scan Bit 0 to Bit 117 |  |  | Device Pin | Boundary Scan Bit 0 to Bit 117 |  |  |
| :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: |
|  | Tristate Control | Output <br> Scan Cell | Input <br> Scan Cell |  | Tristate Control | Output Scan Cell | Input <br> Scan <br> Cell |
| STo7 | 0 | 1 |  | F0i |  |  | 85 |
| STo6 | 2 | 3 |  | STi15 |  |  | 86 |
| STo5 | 4 | 5 |  | STi14 |  |  | 87 |
| STo4 | 6 | 7 |  | STi13 |  |  | 88 |
| STo3 | 8 | 9 |  | STi12 |  |  | 89 |
| STo2 | 10 | 11 |  | STi11 |  |  | 90 |
| STo1 | 12 | 13 |  | STi10 |  |  | 91 |
| STo0 | 14 | 15 |  | STi9 |  |  | 92 |
| ODE |  |  | 16 | STi8 |  |  | 93 |
| CSTo | 17 | 18 |  | STi7 STi6 |  |  | 94 |
| $\overline{\text { DTA }}$ |  | 19 |  | STi5 |  |  | 96 |
| D15 | 20 | 21 |  | STi4 |  |  | 97 |
| D14 | 23 | 24 | 25 | STi3 STi2 |  |  | 98 |
| D13 | 26 | 27 | 28 | STi1 |  |  | 100 |
| D12 | 29 | 30 | 31 | STiO |  |  | 101 |
| D11 | 32 | 33 | $\begin{aligned} & 34 \\ & 37 \end{aligned}$ | STo15 | 102 | 103 |  |
| D10 | 35 | 36 | $37$ | STo14 | 104 | 105 |  |
| D9 | 38 | 39 | $40$ | STo13 | 106 | 107 |  |
| D8 | 41 | 42 |  | STo12 | 108 | 109 |  |
| AD7 | 44 | 45 | 46 | STo11 | 110 | 111 |  |
| AD6 | 47 | 48 | 49 | STo10 | 112 | 113 |  |
| AD5 | 50 | 51 | 52 | STo9 | 114 | 115 |  |
| AD4 | 53 | 54 | 55 | STo8 | 116 | 117 |  |
| AD3 | 56 | 57 | 58 | Table 15 | Boundary | can Regi | Bits |
| AD2 | 59 | 60 | 61 | - | Bund | - |  |
| $\begin{aligned} & \text { AD1 } \\ & \text { AD0 } \end{aligned}$ | $\begin{aligned} & 62 \\ & 65 \end{aligned}$ | $\begin{aligned} & 63 \\ & 66 \end{aligned}$ | $\begin{aligned} & 64 \\ & 67 \end{aligned}$ |  |  |  |  |
| IM |  |  | 68 |  |  |  |  |
| AS/ALE |  |  | 69 |  |  |  |  |
| $\overline{\mathrm{CS}}$ |  |  | 70 |  |  |  |  |
| $\mathrm{R} / \overline{\mathrm{W}} / \overline{\mathrm{WR}}$ |  |  | 71 |  |  |  |  |
| DS/ $\overline{\mathrm{RD}}$ |  |  | 72 |  |  |  |  |
| A7 |  |  | 73 |  |  |  |  |
| A6 |  |  | 74 |  |  |  |  |
| A5 |  |  | 75 |  |  |  |  |
| A4 |  |  | 76 |  |  |  |  |
| A3 |  |  | 77 |  |  |  |  |
| A2 |  |  | 78 |  |  |  |  |
| A1 |  |  | 79 |  |  |  |  |
| A0 |  |  | 80 |  |  |  |  |
| WFPS |  |  | 81 |  |  |  |  |
| RESET |  |  | 82 |  |  |  |  |
| CLK |  |  | 83 |  |  |  |  |
| FE/HCLK |  |  | 84 |  |  |  |  |

## Applications

## Switch Matrix Architectures

The MT90820 is an ideal device for medium to large size switch matrices. Applications where voice and grouped data channels are transported within the same frame, the voice samples have to be time interchanged with a minimum delay while maintaining the integrity of grouped data. To ensure the integrity of grouped data during switching and to provide a minimum delay for voice connections, the MT90820 provides the per-channel selection between variable and constant throughput delay.

This can be selected by the $\overline{\mathrm{V}} / \mathrm{C}$ bit of the Connection Memory. Figure 5 illustrates how four MT90820 devices can be used to form non-blocking switches up to 4096 channels with data rate of $8.192 \mathrm{Mb} / \mathrm{s}$.

## Serial Input Frame Alignment Evaluation

The MT90820 is capable of performing frame alignment evaluation. The frame pulse under evaluation is connected to the FE (frame measure-ment) pin. An external multiplexer is required to selected one of the frame pulses related to the different input streams. Figure 6 gives an example of performing measurement for 16 frame pulses can be performed.


Figure 5 - Switch Matrix with Serial Stream at Various Bit Rates


Note:

1. Use the external mux to select one of the serial frame pulses.
2. To start a measurement cycle, set the Start Frame Evaluation (SFE) bit in the IMS register low for at least 1 frame.
3. Frame evaluation starts when the SFE bit is changed from low to high.
4. Two frames later, the Complete Frame Evaluation (CFE) bit of the Frame Alignment Register (FAR) changes from low to
high to signal the CPU that a valid offset measurement is ready to be read from bit [11:0] of the FAR register.
5. The SFE bit must be set to zero before a new measurement cycle started.

Figure 6 - Serial Input Frame Alignment Evaluation for Various Frame Pulses

## Wide Frame Pulse (WFP) Frame Alignment Mode

When the device is the wide frame pulse mode, the device can operate in the HMVIP and MVIP-90 environment if the input data streams are sampled at $3 / 4$ bit time. When input data stream are sampled at half-bit time as specified in the HMVIP and MVIP-90 standard, the device can only operate with data rate of $2 \mathrm{Mb} / \mathrm{s}$. Refer to the ST-BUS output delay parameter, $\mathrm{t}_{\text {SOD }}$, as specified in the AC Electrical Characteristic table.

The MT90820 is designed to accept a common frame pulse FO i, the 4.096 MHz and 16.384 MHz clocks required by the HMVIP standards. To enable the Width Frame Pulse Frame Alignment Mode, the WFPS pin has to be set to HIGH and the DR1and DRO bits set for $8.192 \mathrm{MB} / \mathrm{s}$ data rate operation.

## Digital Access Cross-Connect System

Figure 7 illustrates how the MT90820 can be used to construct, for example, a $256 \mathrm{E} 1 / \mathrm{T} 1$ digital access cross-connect system (DACS). The system consists of 32 trunk cards each having eight E1 or T1 trunk interfaces for a total of 256 trunks. The central switching block is constructed from 16 MT90820 devices.

Figure 8 shows how an $8,192 \times 8,192$ channel switch can be constructed from $4,096 \times 4,096$ channel
switch modules. Figure 5 shows the implementation of the $4,096 \times 4,096$ channel switch modules from four MT90820 devices. Therefore, 16 MT90820 devices are required to realize an $8,192 \times 8,192$ non-blocking switch module with 64 input and 64 output streams. Each stream has 128 channels per frame for a data rate of $8.192 \mathrm{Mb} / \mathrm{s}$.

Figure 9 shows an eight-stream trunk card block diagram. The MT8986 Multi-rate Digital Switch are used to concentrate the 32 -channel $2.048 \mathrm{Mb} / \mathrm{s}$ ST-BUS (DSTi and DSTo) streams at each E1/T1 trunk onto four 128 -channel $8.192 \mathrm{Mb} / \mathrm{s}$ streams.

It will take 256 MT8986 devices to implement the switching matrix of using 32 -channel $2.048 \mathrm{Mb} / \mathrm{s}$ ST-BUS streams in a square ( $16 \times 16$ ) configuration. A large saving in component cost and board area can be achieved by using 128 -channel $8.192 \mathrm{Mb} / \mathrm{s}$ streams. That is, the same capacity can be achieved using 64 MT8986 devices +16 MT90820 devices for a total of 80 devices.


Figure 7-256 E1/T1 Digital Access Cross-Connect System (DACS)


Figure 8 - 8,192 x 8,192 Channel Switch Matrix


Figure 9 - Trunk Card Block Diagram

## Absolute Maximum Ratings*

|  | Parameter | Symbol | Min | Max | Units |
| :---: | :--- | :---: | :---: | :---: | :---: |
| 1 | Supply Voltage | $\mathrm{V}_{\mathrm{DD}}$ |  | 6.0 | V |
| 2 | Voltage on any pin I/O (other than supply pins) | $\mathrm{V}_{\mathrm{I}}$ | $\mathrm{V}_{\mathrm{SS}}-0.3$ | $\mathrm{~V}_{\mathrm{DD}}+0.3$ | V |
| 3 | Continuous Current at digital outputs | $\mathrm{I}_{0}$ |  | 20 | mA |
| 4 | Package power dissipation (PLCC \& PQFP) | $\mathrm{P}_{\mathrm{D}}$ |  | 2 | W |
| 5 | Storage temperature | $\mathrm{T}_{\mathrm{S}}$ | -65 | +125 | ${ }^{\circ} \mathrm{C}$ |

* Exceeding these values may cause permanent damage. Functional operation under these conditions is not implied.

Recommended Operating Conditions - Voltages are with respect to ground $\left(V_{s s}\right)$ unless otherwise stated.

|  | Characteristics | Sym | Min | Typ | Max | Units | Test Conditions |
| :---: | :--- | :---: | :---: | :---: | :---: | :---: | :---: |
| 1 | Operating Temperature | $\mathrm{T}_{\mathrm{OP}}$ | -40 |  | +85 | ${ }^{\circ} \mathrm{C}$ |  |
| 2 | Positive Supply | $\mathrm{V}_{\mathrm{DD}}$ | 4.75 |  | 5.25 | V |  |
| 3 | Input High Voltage | $\mathrm{V}_{\mathrm{IH}}$ | 2.4 |  | $\mathrm{~V}_{\mathrm{DD}}$ | V | 400 mV noise margin |
| 4 | Input Low Voltage | $\mathrm{V}_{\mathrm{IL}}$ | $\mathrm{V}_{\mathrm{SS}}$ |  | 0.4 | V | 400 mV noise margin |

DC Electrical Characteristics - Voltages are with respect to ground $\left(\mathrm{V}_{\mathrm{ss}}\right)$ unless otherwise stated.

|  |  | Characteristics | Sym | Min | Typ | Max | Units | Test Conditions |
| :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: |
| 1 | INPUTS | Supply Current @ 4 | $I_{\text {D }}$ |  |  | 50 | mA | Output unloaded |
|  |  |  |  |  |  | 90 | mA |  |
|  |  |  |  |  |  | 170 | mA |  |
| 2 |  | Input High Voltage | $\mathrm{V}_{\mathrm{IH}}$ | 2.0 |  |  | V |  |
| 3 |  | Input Low Voltage | $\mathrm{V}_{\text {IL }}$ |  |  | 0.8 | V |  |
| 4 |  | Input Leakage (input pins) <br> Input Leakage (bi-directional pins) | $\begin{aligned} & I_{I L} \\ & I_{\mathrm{BL}} \end{aligned}$ |  |  | $\begin{aligned} & 15 \\ & 50 \end{aligned}$ | $\mu \mathrm{A}$ $\mu \mathrm{A}$ | $0 \leq<\mathrm{V} \leq \mathrm{V}_{\text {DD }}$ See Note 1 |
| 5 |  | Input Pin Capacitance | $\mathrm{C}_{1}$ |  |  | 10 | pF |  |
| 6 | $\bigcirc$ | Output High Voltage | $\mathrm{V}_{\mathrm{OH}}$ | 2.4 |  |  | V | $\mathrm{l}_{\mathrm{OH}}=10 \mathrm{~mA}$ |
| 7 | T | Output Low Voltage | $\mathrm{V}_{\mathrm{OL}}$ |  |  | 0.4 | V | $\mathrm{I}_{\mathrm{OL}}=10 \mathrm{~mA}$ |
| 8 | U | High Impedance Leakage | l OZ |  |  | 5 | $\mu \mathrm{A}$ | $0<\mathrm{V}<\mathrm{V}_{\mathrm{DD}}$ See Note 1 |
| 9 | S | Output Pin Capacitance | $\mathrm{C}_{\mathrm{O}}$ |  |  | 10 | pF |  |

Note:

1. Maximum leakage on pins (output or $\mathrm{I} / \mathrm{O}$ pins in high impedance state) is over an applied voltage ( V )

AC Electrical Characteristics - Timing Parameter Measurement Voltage Levels

|  | Characteristics | Sym | Level | Units | Conditions |
| :---: | :--- | :---: | :---: | :---: | :---: |
| 1 | TTL Threshold | $\mathrm{V}_{\text {TT }}$ | 1.5 | V |  |
| 2 | TTL Rise/Fall Threshold Voltage High | $\mathrm{V}_{\text {HM }}$ | 2.0 | V |  |
| 3 | TTL Rise/Fall Threshold Voltage Low | $\mathrm{V}_{\text {LM }}$ | 0.8 | V |  |

AC Electrical Characteristics - Frame Pulse and CLK

|  | Characteristic | Sym | Min | Typ | Max | Units | Notes |
| :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: |
| 1 | $\begin{aligned} & \text { Frame pulse width (ST-BUS, GCI) } \\ & \text { Bit rate }=2.048 \mathrm{Mb} / \mathrm{s} \\ & \text { Bit rate }=4.096 \mathrm{Mb} / \mathrm{s} \\ & \text { Bit rate }=8.192 \mathrm{Mb} / \mathrm{s} \end{aligned}$ | $\mathrm{t}_{\text {FPW }}$ | $\begin{aligned} & 26 \\ & 26 \\ & 26 \end{aligned}$ |  | $\begin{gathered} 295 \\ 145 \\ 80 \end{gathered}$ | ns ns ns | WFPS Pin = 0 |
| 2 | Frame Pulse Setup time before CLK falling (ST-BUS or GCI) | $\mathrm{t}_{\text {FPS }}$ | 10 |  |  | ns | WFPS Pin = 0 |
| 3 | Frame Pulse Hold Time from CLK falling (ST-BUS or GCI) | $\mathrm{t}_{\text {FPH }}$ | 16 |  |  | ns | WFPS Pin = 0 |
| 4 | $\begin{aligned} & \text { CLK Period } \\ & \text { Bit rate }=2.048 \mathrm{Mb} / \mathrm{s} \\ & \text { Bit rate }=4.096 \mathrm{Mb} / \mathrm{s} \\ & \text { Bit rate }=8.192 \mathrm{Mb} / \mathrm{s} \end{aligned}$ | $\mathrm{t}_{\mathrm{CP}}$ | $\begin{gathered} 190 \\ 110 \\ 55 \end{gathered}$ |  | $\begin{gathered} 300 \\ 150 \\ 70 \end{gathered}$ | ns ns ns | WFPS Pin = 0 |
| 5 | CLK Pulse Width High <br> Bit rate $=2.048 \mathrm{Mb} / \mathrm{s}$ <br> Bit rate $=4.096 \mathrm{Mb} / \mathrm{s}$ <br> Bit rate $=8.192 \mathrm{Mb} / \mathrm{s}$ | $\mathrm{t}_{\mathrm{CH}}$ | $\begin{aligned} & 85 \\ & 50 \\ & 20 \\ & \hline \end{aligned}$ |  | $\begin{gathered} 150 \\ 75 \\ 40 \\ \hline \end{gathered}$ |  | WFPS Pin = 0 |
| 6 | $\begin{aligned} \hline \text { CLK Pulse Width Low } \\ \text { Bit rate }=2.048 \mathrm{Mb} / \mathrm{s} \\ \text { Bit rate }=4.096 \mathrm{Mb} / \mathrm{s} \\ \text { Bit rate }=8.192 \mathrm{Mb} / \mathrm{s} \end{aligned}$ | $\mathrm{t}_{\mathrm{CL}}$ | $\begin{aligned} & 85 \\ & 50 \\ & 20 \\ & \hline \end{aligned}$ |  | $\begin{gathered} 150 \\ 75 \\ 40 \\ \hline \end{gathered}$ | $\begin{aligned} & \text { ns } \\ & \text { ns } \end{aligned}$ ns | WFPS Pin = 0 |
| 7 | Clock Rise/Fall Time | $\mathrm{t}_{\mathrm{r}}, \mathrm{t}_{\mathrm{f}}$ |  |  | 10 | ns |  |
| 8 | Wide frame pulse width Bit rate $=8.192 \mathrm{Mb} / \mathrm{s}$ | $\mathrm{t}_{\text {HFPW }}$ | 195 |  | 295 | ns | WFPS Pin = 1 |
| 9 | Frame Pulse Setup Time before HCLK falling | $\mathrm{t}_{\text {HFPS }}$ | 10 |  | 150 | ns | WFPS Pin = 1 |
| 10 | Frame Pulse Hold Time from HCLK falling | $\mathrm{t}_{\text {HFPH }}$ | 20 |  | 150 | ns | WFPS Pin = 1 |
| 11 | HCLK (4.096MHz) Period Bit rate $=8.192 \mathrm{Mb} / \mathrm{s}$ | $\mathrm{t}_{\mathrm{HCP}}$ | 190 |  | 300 | ns | WFPS Pin = 1 |
| 12 | HCLK (4.096MHz) Pulse Width High Bit rate $=8.192 \mathrm{Mb} / \mathrm{s}$ | $\mathrm{t}_{\mathrm{HCH}}$ | 85 |  | 150 | ns | WFPS Pin = 1 |
| 13 | HCLK (4.096MHz) Pulse Width Low Bit rate $=8.192 \mathrm{Mb} / \mathrm{s}$ | $\mathrm{t}_{\mathrm{HCL}}$ | 85 |  | 150 | ns | WFPS Pin = 1 |
| 14 | HCLK Rise/Fall Time | $\mathrm{t}_{\mathrm{Hr}}, \mathrm{t}_{\mathrm{Hf}}$ |  |  | 10 | ns |  |
| 15 | Delay between falling edge of HCLK and falling edge of CLK | $\mathrm{t}_{\text {DIF }}$ | -10 |  | 10 | ns | WFPS Pin $=0$ or 1 |

## AC Electrical Characteristics - Serial Streams for ST-BUS and GCI Backplanes

|  | Characteristic | Sym | Min | Typ | Max | Units | Test Conditions |
| :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: |
| 1 | Sti Set-up Time | $\mathrm{t}_{\text {SIS }}$ | 0 |  |  | ns |  |
| 2 | Sti Hold Time | $\mathrm{t}_{\text {SIH }}$ | 20 |  |  | ns |  |
| 3 | Sto Delay - Active to Active <br> @ $2.048 \mathrm{Mb} / \mathrm{s}$ mode <br> @ $4.096 \mathrm{Mb} / \mathrm{s}$ mode <br> @ $8.192 \mathrm{Mb} / \mathrm{s}$ mode <br> @ $8.192 \mathrm{Mb} / \mathrm{s}$ mode | $\mathrm{t}_{\text {SOD }}$ |  |  | $\begin{aligned} & 58 \\ & 58 \\ & 58 \\ & 39 \end{aligned}$ | ns <br> ns <br> ns ns | $\begin{aligned} & \mathrm{C}_{\mathrm{L}}=200 \mathrm{pF} \\ & \mathrm{C}_{\mathrm{L}}=200 \mathrm{pF} \\ & \mathrm{C}_{\mathrm{L}}=200 \mathrm{pF} \\ & \mathrm{C}_{\mathrm{L}}=30 \mathrm{pF} \end{aligned}$ |
| 4 | STo delay - Active to High-Z $2.048 \mathrm{Mb} / \mathrm{s}$ mode <br> $4.096 \mathrm{Mb} / \mathrm{s}$ mode <br> $8.192 \mathrm{Mb} / \mathrm{s}$ mode | $t_{D Z}$ |  |  | $\begin{aligned} & 37 \\ & 37 \\ & 37 \end{aligned}$ | $\begin{aligned} & \text { ns } \\ & \text { ns } \\ & \text { ns } \end{aligned}$ | $R_{L}=1 \mathrm{~K}, \mathrm{C}_{\mathrm{L}}=200 \mathrm{pF}$, See Note 1 $R_{L}=1 K, C_{L}=200 p F$, See Note 1 $R_{L}=1 K, C_{L}=200 p F$, See Note 1 |
| 5 | Sto delay - High-Z to Active $2.048 \mathrm{Mb} / \mathrm{s}$ mode $4.096 \mathrm{Mb} / \mathrm{s}$ mode 8.192 Mb/s mode | ${ }^{\text {t }}$ D |  |  | $\begin{aligned} & 37 \\ & 37 \\ & 37 \end{aligned}$ | $\begin{aligned} & \text { ns } \\ & \text { ns } \\ & \text { ns } \end{aligned}$ | $\begin{aligned} & R_{L}=1 \mathrm{~K}, C_{L}=200 \mathrm{pF} \text {, See Note } 1 \\ & R_{\mathrm{L}}=1 \mathrm{~K}, C_{L}=200 \mathrm{pF} \text {, See Note } 1 \\ & R_{L}=1 \mathrm{~K}, C_{L}=200 \mathrm{pF} \text {, See Note } 1 \end{aligned}$ |
| 6 | Output Driver Enable (ODE) Delay <br> $2.048 \mathrm{Mb} / \mathrm{s}$ mode <br> $4.096 \mathrm{Mb} / \mathrm{s}$ mode <br> $8.192 \mathrm{Mb} / \mathrm{s}$ mode | $t_{\text {ODE }}$ |  |  | $\begin{aligned} & 37 \\ & 37 \\ & 37 \end{aligned}$ | $\begin{aligned} & \text { ns } \\ & \text { ns } \\ & \text { ns } \end{aligned}$ | $R_{L}=1 \mathrm{~K}, \mathrm{C}_{\mathrm{L}}=200 \mathrm{pF}$, See Note 1 $R_{L}=1 K, C_{L}=200 p F$, See Note 1 $R_{L}=1 K, C_{L}=200 p F$, See Note 1 |
| 7 | CSTo Output Delay <br> $2.048 \mathrm{Mb} / \mathrm{s}$ mode <br> $4.096 \mathrm{Mb} / \mathrm{s}$ mode <br> $8.192 \mathrm{Mb} / \mathrm{s}$ mode | ${ }^{\text {XCD }}$ |  |  | $\begin{aligned} & 58 \\ & 58 \\ & 58 \end{aligned}$ |  | $\begin{aligned} & \mathrm{C}_{\mathrm{L}}=200 \mathrm{pF} \\ & \mathrm{C}_{\mathrm{L}}=200 \mathrm{pF} \\ & \mathrm{C}_{\mathrm{L}}=200 \mathrm{pF} \\ & \hline \end{aligned}$ |

Note:

1. High Impedance is measured by pulling to the appropriate rail with $R_{L}$, with timing corrected to cancel time taken to discharge $C_{L}$.


Note 1:
$2.048 \mathrm{Mb} / \mathrm{s}$ mode, last channel = ch 31,
$4.196 \mathrm{Mb} / \mathrm{s}$ mode, last channel $=$ ch 63 ,
$8.192 \mathrm{Mb} /$ s mode, last channel $=$ ch 127.

Figure 11 - ST-BUS Timing for $2.048 \mathrm{Mb} / \mathrm{s}$ and High Speed Serial Interface at $4.096 \mathrm{Mb} / \mathrm{s}$ or 8.192 Mb/s, when WFPS pin = 0 .

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Note 1:
$2 \mathrm{Mb} / \mathrm{s}$ mode, last channel $=$ ch 31 ,
$4 \mathrm{Mb} / \mathrm{s}$ mode, last channel $=$ ch 63 ,
$8 \mathrm{Mb} / \mathrm{s}$ mode, last channel $=$ ch 127

Figure 12 - GCl Timing at $2.048 \mathrm{Mb} / \mathrm{s}$ and High Speed Serial Interface at 4.096 Mb/s or 8.192 Mb/s, when WFPS pin = 0


Figure 13 - WFP Bus Timing for High Speed Serial Interface (8.192Mb/s), when WFPS pin = 1 Note:

1. High Impedance is measured by pulling to the appropriate rail with $R_{L}$, with timing corrected to cancel time taken to discharge $C_{L}$.


Figure 14-Serial Output and External Control


Figure 15 - Output Driver Enable (ODE)

## AC Electrical Characteristics - Multiplexed Bus Timing (Mode 1)

|  | Characteristics | Sym | Min | Typ | Max | Units | Test Conditions |
| :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: |
| 1 | ALE pulse width | $\mathrm{t}_{\text {ALW }}$ | 20 |  |  | ns |  |
| 2 | Address setup from ALE falling | $\mathrm{t}_{\text {ADS }}$ | 10 |  |  | ns |  |
| 3 | Address hold from ALE falling | $\mathrm{t}_{\text {ADH }}$ | 10 |  |  | ns |  |
| 4 | $\overline{\mathrm{RD}}$ active after ALE falling | $\mathrm{t}_{\text {ALRD }}$ | 10 |  |  | ns |  |
| 5 | Data setup from DTA Low on Read | $\mathrm{t}_{\text {DDR }}$ | 10 |  |  | ns | $\mathrm{C}_{\mathrm{L}}=150 \mathrm{pF}$ |
| 6 | $\overline{\mathrm{CS}}$ hold after $\overline{\mathrm{RD}} / \overline{\mathrm{WR}}$ | $\mathrm{t}_{\text {CSRW }}$ | 0 |  |  | ns |  |
| 7 | $\overline{\mathrm{RD}}$ pulse width (fast read) | $\mathrm{t}_{\text {RW }}$ |  | 80 |  | ns |  |
| 8 | $\overline{\mathrm{CS}}$ setup from $\overline{\mathrm{RD}}$ | $\mathrm{t}_{\mathrm{CSR}}$ | 0 |  |  | ns |  |
| 9 | Data hold after $\overline{\mathrm{RD}}$ | $\mathrm{t}_{\mathrm{DHR}}$ | 10 |  | 75 | ns | $\mathrm{C}_{\mathrm{L}}=150 \mathrm{pF}, \mathrm{R}_{\mathrm{L}}=1 \mathrm{~K},$ <br> Note 1. |
| 10 | $\overline{\text { WR pulse width (fast write) }}$ | $t_{\text {WW }}$ |  | 90 |  | ns |  |
| 11 | $\overline{\text { WR delay after ALE falling }}$ | $\mathrm{t}_{\text {ALWR }}$ | 10 |  |  | ns |  |
| 12 | $\overline{\mathrm{CS}}$ setup from $\overline{\mathrm{WR}}$ | $\mathrm{t}_{\text {csw }}$ | 0 |  |  | ns |  |
| 13 | Data setup from $\overline{\mathrm{WR}}$ (fast write) | $\mathrm{t}_{\text {DSW }}$ | 90 |  |  | ns |  |
| 14 | Valid Data Delay on write (slow write) | $\mathrm{t}_{\text {SWD }}$ |  |  | 122 | ns |  |
| 15 | Data hold after $\overline{W R}$ inactive | $t_{\text {DHW }}$ | 10 |  |  | ns |  |
| 16 | Acknowledgment Delay: Reading/Writing Registers Reading/Writing Memory, @ 2Mb/s @ 4Mb/s @ 8Mb/s | $\mathrm{t}_{\text {AKD }}$ |  |  | $\begin{gathered} 55 / 60 \\ 760 / 780 \\ 400 / 420 \\ 220 / 240 \\ \hline \end{gathered}$ | ns ns ns ns | $\begin{aligned} & \mathrm{C}_{\mathrm{L}}=150 \mathrm{pF} \\ & \mathrm{C}_{\mathrm{L}}=150 \mathrm{pF} \\ & \mathrm{C}_{\mathrm{L}}=150 \mathrm{pF} \\ & \mathrm{C}_{\mathrm{L}}=150 \mathrm{pF} \end{aligned}$ |
| 17 | Acknowledgment Hold Time | $\mathrm{t}_{\text {AKH }}$ |  | 45 | 80 | ns | $\mathrm{C}_{\mathrm{L}}=150 \mathrm{pF}, \mathrm{R}_{\mathrm{L}}=1 \mathrm{~K},$ <br> Note 1. |

Note:

1. High Impedance is measured by pulling to the appropriate rail with $R_{L}$, with timing corrected to cancel time taken to discharge $C_{L}$.


Figure 16 - Multiplexed Bus Timing (Mode 1)

## AC Electrical Characteristics - Multiplexed Bus Timing (Mode 2)

|  | Characteristics | Sym | Min | Typ | Max | Units | Test Conditions |
| :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: |
| 1 | AS pulse width | $\mathrm{t}_{\text {ASW }}$ | 80 |  |  | ns |  |
| 2 | Address setup from AS falling | $\mathrm{t}_{\text {ADS }}$ | 10 |  |  | ns |  |
| 3 | Address hold from AS falling | $\mathrm{t}_{\text {ADH }}$ | 10 |  |  | ns |  |
| 4 | Data setup from DTA Low on Read | $\mathrm{t}_{\text {DDR }}$ | 10 |  |  | ns | $\mathrm{C}_{\mathrm{L}}=150 \mathrm{pF}$ |
| 5 | $\overline{\mathrm{CS}}$ hold after DS falling | $\mathrm{t}_{\mathrm{CSH}}$ | 0 |  |  | ns |  |
| 6 | $\overline{\mathrm{CS}}$ setup from DS rising | $\mathrm{t}_{\text {CSS }}$ | 0 |  |  | ns |  |
| 7 | Data hold after write | $t_{\text {DHW }}$ | 10 |  |  | ns |  |
| 8 | Data setup from DS -Write (fast write) | $t_{\text {DWS }}$ | 25 |  |  | ns |  |
| 9 | Valid Data Delay on write (slow write) | $t_{\text {SWD }}$ |  |  | 122 | ns |  |
| 10 | R/ $/ \mathrm{W}$ setup from DS rising | $t_{\text {RWS }}$ | 60 |  |  | ns |  |
| 11 | R//W hold after DS falling | $\mathrm{t}_{\text {RWH }}$ | 10 |  |  | ns |  |
| 12 | Data hold after read | $t_{\text {DHR }}$ | 10 | 50 | 75 | ns | $\mathrm{C}_{\mathrm{L}}=150 \mathrm{pF}, \mathrm{R}_{\mathrm{L}}=1 \mathrm{~K},$ <br> Note 1 |
| 13 | DS delay after AS falling | $t_{\text {DSH }}$ | 10 |  |  | ns |  |
| 14 | Acknowledgment Delay: Reading/Writing Registers Reading/Writing Memory, @ 2Mb/s @ 4Mb/s @ $8 \mathrm{Mb} / \mathrm{s}$ | $\mathrm{t}_{\text {AKD }}$ |  |  | $\begin{gathered} 55 / 60 \\ 760 / 780 \\ 400 / 420 \\ 220 / 240 \end{gathered}$ | $\begin{aligned} & \text { ns } \\ & \text { ns } \\ & \text { ns } \\ & \text { ns } \end{aligned}$ | $\begin{aligned} & C_{\mathrm{L}}=150 \mathrm{pF} \\ & \mathrm{C}_{\mathrm{L}}=150 \mathrm{pF} \\ & \mathrm{C}_{\mathrm{L}}=150 \mathrm{pF} \\ & \mathrm{C}_{\mathrm{L}}=150 \mathrm{pF} \end{aligned}$ |
| 15 | Acknowledgment Hold Time | $\mathrm{t}_{\text {AKH }}$ |  | 45 | 80 | ns | $\mathrm{C}_{\mathrm{L}}=150 \mathrm{pF}, \mathrm{R}_{\mathrm{L}}=1 \mathrm{~K},$ <br> Note 1 |

Note 1. High Impedance is measured by pulling to the appropriate rail with $\mathrm{R}_{\mathrm{L}}$, with timing corrected to cancel time taken to discharge $\mathrm{C}_{\mathrm{L}}$.


## AC Electrical Characteristics - Motorola Non-Multiplexed Bus Mode

|  | Characteristics | Sym | Min | Typ | Max | Units | Test Conditions |
| :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: |
| 1 | $\overline{\mathrm{CS}}$ setup from DS falling | $\mathrm{t}_{\mathrm{CSS}}$ | 0 |  |  | ns |  |
| 2 | R/ $\bar{W}$ setup from DS falling | $t_{\text {RWS }}$ | 10 |  |  | ns |  |
| 3 | Address setup from DS falling | $\mathrm{t}_{\text {ADS }}$ | 2 |  |  | ns |  |
| 4 | $\overline{\mathrm{CS}}$ hold after DS rising | $\mathrm{t}_{\mathrm{CSH}}$ | 0 |  |  | ns |  |
| 5 | R//W hold after DS rising | $\mathrm{t}_{\text {RWH }}$ | 5 |  |  | ns |  |
| 6 | Address hold after DS rising | $\mathrm{t}_{\text {ADH }}$ | 5 |  |  | ns |  |
| 7 | Data setup from DTA Low on Read | $\mathrm{t}_{\text {DDR }}$ | 0 |  |  | ns | $\mathrm{C}_{\mathrm{L}}=150 \mathrm{pF}$ |
| 8 | Data hold on read | $\mathrm{t}_{\text {DHR }}$ | 10 | 50 | 75 | ns | $\mathrm{C}_{\mathrm{L}}=150 \mathrm{pF}, \mathrm{R}_{\mathrm{L}}=1 \mathrm{~K}$ <br> Note 1 |
| 9 | Data setup on write (fast write) | $\mathrm{t}_{\text {DSW }}$ | 20 |  |  | ns |  |
| 10 | Valid Data Delay on write (slow write) | tSWD |  |  | 122 | ns |  |
| 11 | Data hold on write | $\mathrm{t}_{\text {DHW }}$ | 8 |  |  | ns |  |
| 12 | Acknowledgment Delay: Reading/Writing Registers Reading/Writing Memory, @ 2Mb/s @ 4Mb/s @ 8Mb/s | $\mathrm{t}_{\text {AKD }}$ |  |  | $\begin{gathered} 55 / 60 \\ 760 / 780 \\ 400 / 420 \\ 220 / 240 \end{gathered}$ | ns <br> ns <br> ns <br> ns | $\begin{aligned} & C_{L}=150 \mathrm{pF} \\ & C_{L}=150 \mathrm{pF} \\ & C_{\mathrm{L}}=150 \mathrm{pF} \\ & C_{\mathrm{L}}=150 \mathrm{pF} \end{aligned}$ |
| 13 | Acknowledgment Hold Time | $\mathrm{t}_{\text {AKH }}$ |  | 45 | 80 | ns | $\mathrm{C}_{\mathrm{L}}=150 \mathrm{pF}, \mathrm{R}_{\mathrm{L}}=1 \mathrm{~K},$ <br> Note 1 |

Note:

1. High Impedance is measured by pulling to the appropriate rail with $R_{L}$, with timing corrected to cancel time taken to discharge $C_{L}$.


Figure 18 - Motorola Non-Multiplexed Bus Timing

## MT90820 cMOS

Notes:

